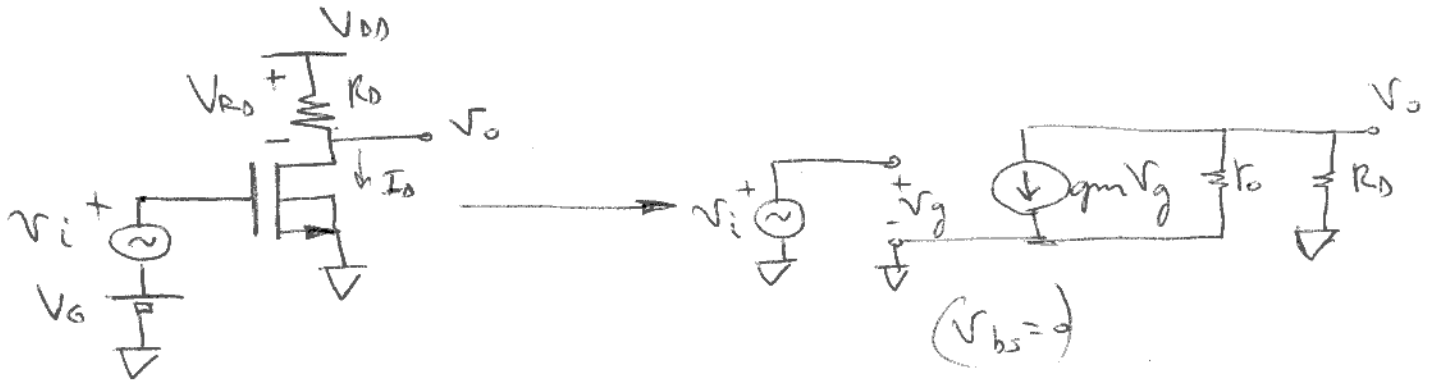


SMALL-SIGNAL ANALYSIS OF CS. AMPLIFIER

3



$$\frac{v_o}{v_i} = -g_m (r_o \parallel R_D) \rightarrow \text{ASSUME } r_o \gg R_D$$

$$\frac{v_o}{v_i} \approx -g_m R_D = -g_m \frac{V_{RD}}{I_D}$$

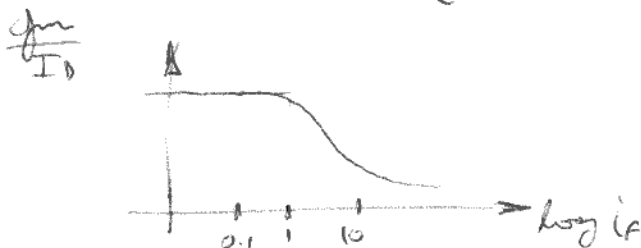
- $V_{RD} = V_{DD} - V_{DS} \rightarrow$ MAXIMIZED WHEN $V_{DS} = V_{DSAT}$
- OTHER THAN V_{RD} , MUST FIND OPTIMUM $\frac{g_m}{I_D}$

$$\frac{g_m}{I_D} = \frac{\frac{2I_S}{nV_T} (\sqrt{1+if} - 1)}{I_S if} = \frac{2}{nV_T} \frac{(1+if) - 1}{if(\sqrt{1+if} + 1)} = \frac{2}{nV_T} \frac{1}{(\sqrt{1+if} + 1)}$$

W.I) $\frac{g_m}{I_D} \approx \frac{1}{nV_T}$ CONSTANT

S.I) $\frac{g_m}{I_D} \approx \frac{2}{nV_T \sqrt{if}}$ DECREASES WITH if

\therefore MAX GAIN @ WI. $\left\{ \begin{array}{l} \frac{g_m}{I_D} \text{ MAX} \\ V_{RD} \text{ MAX SINCE } V_{DSAT} \text{ MIN} \end{array} \right.$



EXAMPLE : 1.8V SUPPLY

$$I_{SQ} = 300\text{ nA}, \quad n = 1.25$$

$$\frac{W}{L} = 20$$

a) CALCULATE MAX GAIN

$$\left. \begin{aligned} V_{RD \text{ max}} &\approx 1.8\text{V} - 100\text{mV} = 1.7\text{V} \\ \frac{g_m}{I_{D \text{ max}}} &\approx \frac{1}{1.25 \times 26\text{mV}} = 30.76 \text{ V}^{-1} \end{aligned} \right\} |a_{V \text{ max}}| = 52.3$$

(NOT MUCH)

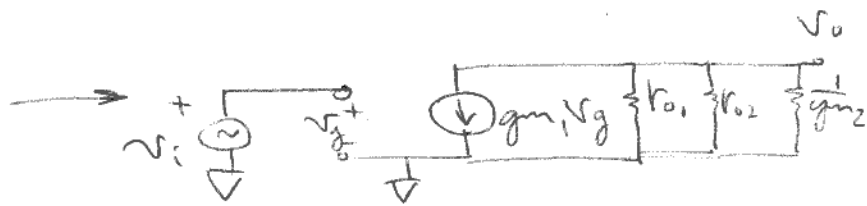
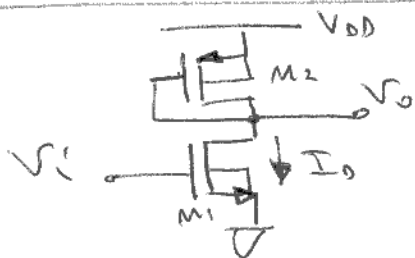
b) CALCULATE R_D :

FOR W.I., $i_f \approx 0.1$ (ASSUME)

$$I_D = 300\text{ nA} \times 20 \times 0.1 = 600\text{ nA}$$

$$R_D = \frac{1.7\text{V}}{600\text{ nA}} = 2.83 \text{ M}\Omega \quad \text{VERY HIGH}$$

DIODE-CONNECTED LOAD



$$\frac{v_o}{v_i} = -g_{m1} (r_{o1} \parallel r_{o2} \parallel \frac{1}{g_{m2}}) \approx -\frac{g_{m1}}{g_{m2}} = \frac{\frac{2 I_{S1}}{n_1 V_T} (\sqrt{1+i_{f1}} - 1)}{\frac{2 I_{S2}}{n_2 V_T} (\sqrt{1+i_{f2}} - 1)}$$

$$\frac{v_o}{v_i} \approx \frac{n_1 \mu_1 (\frac{W}{L})_1 (\sqrt{1+i_{f1}} - 1)}{n_2 \mu_2 (\frac{W}{L})_2 (\sqrt{1+i_{f2}} - 1)} \rightarrow \text{W.I.} \quad \frac{n_1 \mu_1 (\frac{W}{L})_1 (i_{f1})}{n_2 \mu_2 (\frac{W}{L})_2 (i_{f2})}$$

$$\rightarrow \text{S.I.} \quad \frac{n_1 \mu_1 (\frac{W}{L})_1}{n_2 \mu_2 (\frac{W}{L})_2} \frac{\sqrt{i_{f1}}}{\sqrt{i_{f2}}}$$

BETTER:

$$\frac{g_{m1}}{g_{m2}} = \frac{g_{m1}}{I_D} \cdot \frac{I_D}{g_{m2}}$$

$$\frac{v_o}{v_i} \approx - \frac{2}{n_1} \frac{(\sqrt{1+i_{f1}} - 1)}{i_{f1}} \times \frac{n_2 i_{f2}}{2(\sqrt{1+i_{f2}} - 1)}$$

$$\frac{v_o}{v_i} = - \frac{n_2}{n_1} \frac{i_{f2}}{i_{f1}} \frac{(\sqrt{1+i_{f1}} - 1)}{(\sqrt{1+i_{f2}} - 1)}$$

W.I) $\frac{v_o}{v_i} \approx - \frac{n_2}{n_1} \approx 1$ NO GAIN

S.I) $\frac{v_o}{v_i} \approx - \frac{n_2}{n_1} \sqrt{\frac{i_{f2}}{i_{f1}}} = - \frac{n_2}{n_1} \sqrt{\frac{\mu_n n_1 (\frac{W}{L})_1}{\mu_p n_2 (\frac{W}{L})_2}}$

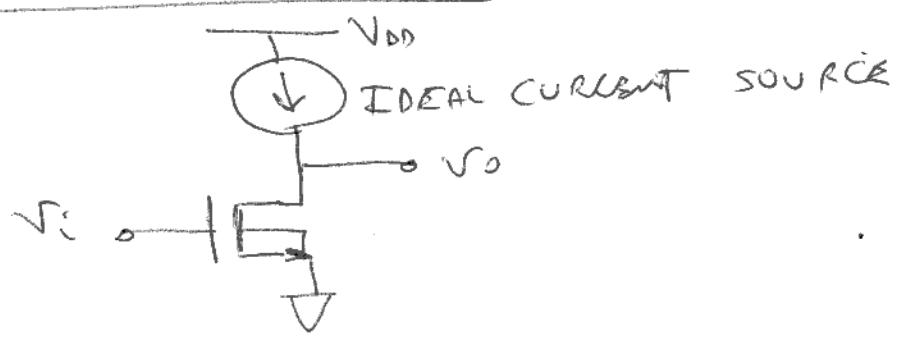
$$= - \sqrt{\frac{\mu_n n_2 (\frac{W}{L})_1}{\mu_p n_1 (\frac{W}{L})_2}}$$

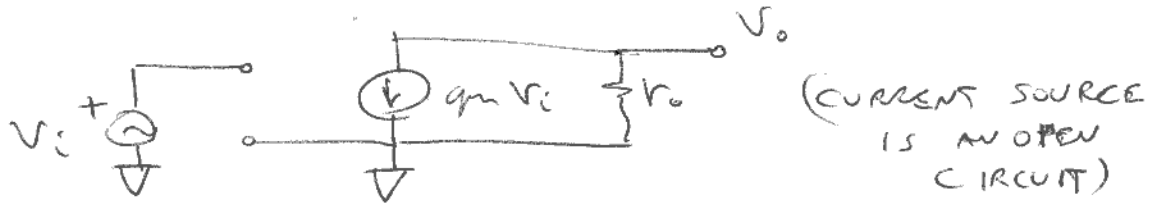
→ GAIN INCREASES IF
 $i_{f2} \gg i_{f1}$
 $(\frac{W}{L})_1 \gg (\frac{W}{L})_2$
 (BUT GAIN SMALL DUE TO $\sqrt{\quad}$)

EXAMPLE: $\mu_n = 400 \frac{cm^2}{Vs}$
 $\mu_p = 160 \frac{cm^2}{Vs}$
 $W_n = 100 \mu m$
 $W_p = 10 \mu m$
 $L_n = L_p, n_1 = n_2$

$$\frac{v_o}{v_i} \approx - 5 \rightarrow i_{f2} \approx 25 i_{f1}$$

INTRINSIC GAIN STAGE





$$\frac{V_o}{V_i} = -g_m r_o = -g_m \frac{V_A}{I_D} = -\frac{g_m}{I_D} V_A$$

∴ TO MAXIMIZE GAIN $\left\{ \begin{array}{l} \text{MAXIMIZE } \frac{g_m}{I_D} \text{ (W.I.)} \\ \text{MAXIMIZE } V_A \text{ (LONG CHANNEL)} \end{array} \right.$

• PROBLEM WITH THAT: FREQUENCY RESPONSE IS BAD.

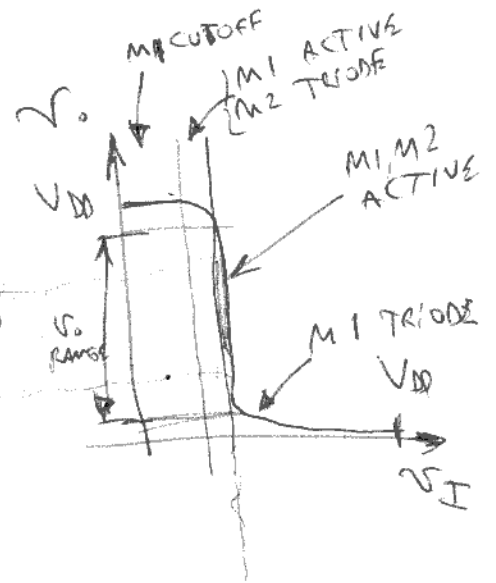
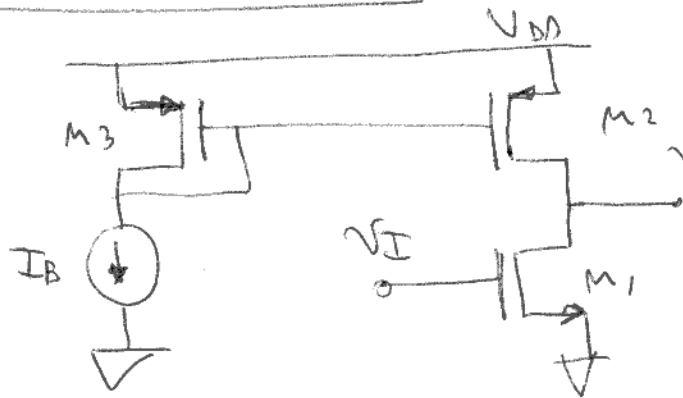
$$V_A = \frac{L}{\left| \frac{\partial L}{\partial V_{DS}} \right|}$$

EXAMPLE: $\left| \frac{\partial L}{\partial V_{DS}} \right| = 0.02 \frac{\mu m}{V}$, $L = 2 \mu m$, $n = 1.2$

• CALCULATE MAXIMUM INTRINSIC GAIN:

$$\left. \begin{array}{l} \frac{g_m}{I_D} \Big|_{\max} \approx 32.05 \text{ V}^{-1} \\ V_A = 100 \text{ V} \end{array} \right\} \left| \frac{V_o}{V_i} \right|_{\max} = 3205 \rightarrow \text{IN ONE STAGE}$$

CURRENT SOURCE LOAD



• V_o RANGE FOR LARGE GAIN:

$$\underbrace{V_{DS1SAT}}_{V_T \left(\sqrt{1 + \beta_1} + 3 \right)} \leq V_o \leq V_{DD} - \underbrace{V_{DS2SAT}}_{V_T \left(\sqrt{1 + \beta_2} + 3 \right)}$$

$$V_T \left(\sqrt{1 + \frac{I_B}{I_{S1}}} + 3 \right) \leq V_o \leq V_{DD} - V_T \left(\sqrt{1 + \frac{I_B}{I_{S2}}} + 3 \right)$$

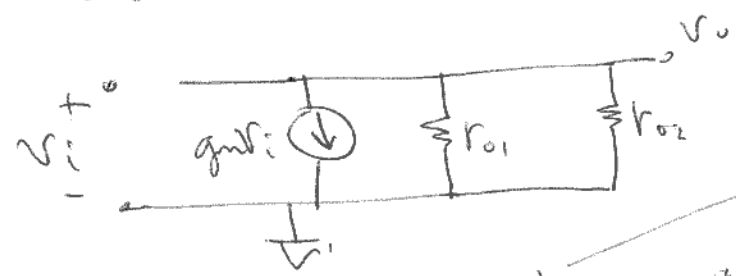
• INPUT VOLTAGE FOR ACTIVE REGION $\Rightarrow I_{D1} = I_B$

$$I_B = I_{S1} \beta_1 \quad (i_{fi} \text{ NEGLIGIBLE})$$

$$V_I = ? \quad \frac{V_I - V_{th}}{n} - \underbrace{V_S}_{=0} = V_T \mathcal{F}(i_{fi})$$

$$\boxed{V_I = n V_T \mathcal{F}\left(\frac{I_B}{\beta_1}\right) + V_{th}}$$

• SMALL-SIGNAL ANALYSIS:



OBSERVATION:
 LARGE $I_D \Rightarrow \begin{cases} g_m \uparrow \\ r_o \downarrow \end{cases}$
 SMALL $I_D \Rightarrow \begin{cases} g_m \downarrow \\ r_o \uparrow \end{cases}$

$$\frac{v_o}{v_i} = -g_m (r_{o1} \parallel r_{o2}) = -\frac{g_m}{(g_{ds1} + g_{ds2})}$$

$$g_{ds1} = \frac{I_{D1} \beta_1}{V_{A1}} \quad g_{ds2} = \frac{I_{D2}}{V_{A2}} \Rightarrow \frac{v_o}{v_i} = -\frac{g_m}{I_D} \cdot \frac{1}{\left(\frac{1}{V_{A1}} + \frac{1}{V_{A2}}\right)}$$

$$= \frac{2}{n V_T} \frac{1}{\left(\frac{1}{V_{A1}} + \frac{1}{V_{A2}}\right) \left(\sqrt{1 + \beta} + 1\right)}$$

EXAMPLE:

$$\left. \begin{array}{l} V_{A1} = 50V \\ V_{A2} = 30V \\ \beta_1 = 10 \end{array} \right\} \frac{v_o}{v_i} = -278$$

$$\beta_1 = 1 \Rightarrow \frac{v_o}{v_i} = -497$$

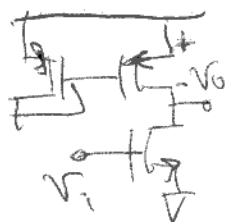
$$\beta_1 = 0.01 \Rightarrow \frac{v_o}{v_i} = -586$$

$$\beta_1 = 100 \Rightarrow -108$$

EXAMPLE

- DESIGN A COMMON-SOURCE AMPLIFIER WITH A GAIN OF AT LEAST 150, SLEW RATE @ OUTPUT $\geq 10 \frac{V}{\mu s}$ WITH $C_L = 1 pF$.

$\left| \frac{\partial L}{\partial V_{DS}} \right|_n = 0.03 \frac{\mu m}{V}$ $I_{SQn} = 280 nA$ $V_{thn} = 0.5V$ $n = 1.25$
 $\left| \frac{\partial L}{\partial V_{DS}} \right|_p = 0.04 \frac{\mu m}{V}$ $I_{SQp} = 80 nA$ $V_{thp} = 0.55V$ $n = 1.25$



- $L \geq 1 \mu m$ FOR ALL TRANSISTORS (MATCHING REQUIREMENT)
- TRY TO USE LOW AREA. $V_{DD} = 1.5V$

SOLUTION: $\left| \frac{V_O}{V_i} \right| = \frac{2}{n V_T} \frac{1}{\left(\frac{1}{\beta_{A1}} + \frac{1}{\beta_{A2}} \right) \left(\sqrt{1 + \beta_{F1}} + 1 \right)} = 150$

- ASSUME $L = 1 \mu m$ TO SAVE AREA $\Rightarrow \begin{cases} V_{A1} = 33.3V \\ V_{A2} = 25V \end{cases}$

$\therefore \beta_{F1} \leq 2.73$

~~TAKE $\beta_{F1} = 2$~~

$I_B \geq C_L \frac{dV_O}{dt} = 10 \mu A$

$\frac{dV_O}{dt} = \frac{I_B}{C_L}$

(BJT: $30 fA/s$)

- TAKE $\beta_{F1} = 2$ (SAVE AREA), $I_B = 12 \mu A$

$\therefore \left(\frac{W}{L} \right)_1 = \frac{12 \mu A}{2 \times 280 nA} = 21.4 \approx 21.5$

$\begin{cases} W_1 = 21.5 \mu m \\ L_1 = 1 \mu m \end{cases}$

$g_{m1} = 0.271 mS$
 $r_{O1} = 33 M\Omega$
 $r_{O2} = 2.5 M\Omega$
 $\frac{1}{g_{m1}} = 3.6 k\Omega \ll r_{O1}$

- β_{F2} NOT IMPORTANT FOR GAIN BUT IMPORTANT FOR V_O RANGE

V_O RANGE

TAKE $\begin{cases} W_2 = 10 \mu m \\ L_2 = 1 \mu m \end{cases} \Rightarrow \left| V_{DSAT2} \right| \approx V_T \left(\sqrt{1 + \frac{R_{M1}}{\beta_{F2} I_{SQ2}}} + 3 \right) = 0.182V$

$V_{DSAT1} = 123 mV$

$123 mV \leq V_O \leq 1.318V$